



From DC up to 30 GHz “Real-Time” oscilloscope Acquisition

“Advanced Measurements ...not only Signal Integrity ” - July2009



Agenda

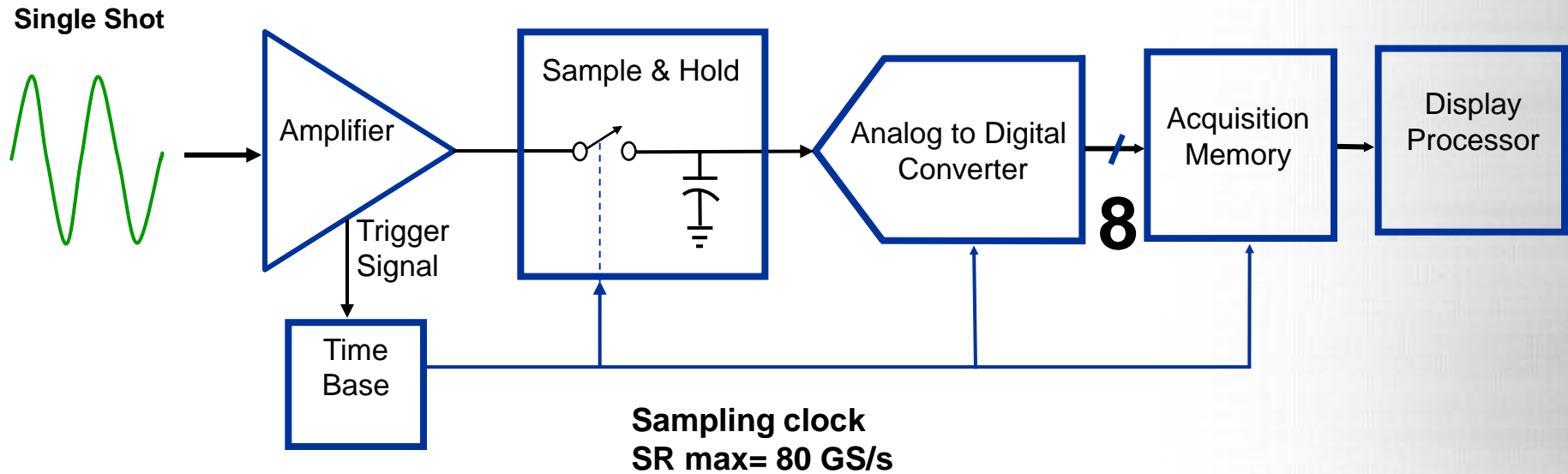
- ❑ **How to achieve 30 GHz BW and 80 GS/s SR ?**
 - ❑ front-end architecture
 - ❑ DBI technology
- ❑ **Frequency Response Optimization**
 - ❑ Flatness
 - ❑ Pulse Response
 - ❑ Eye Response
- ❑ **Time and Frequency Domain Analysis**
 - ❑ Advanced FFT and Spectrum Analysis
- ❑ **Emulating / De-embedding / Processing**
 - ❑ Cable de-embedding
 - ❑ Using the S-parameters
 - ❑ User-defined custom algorithms

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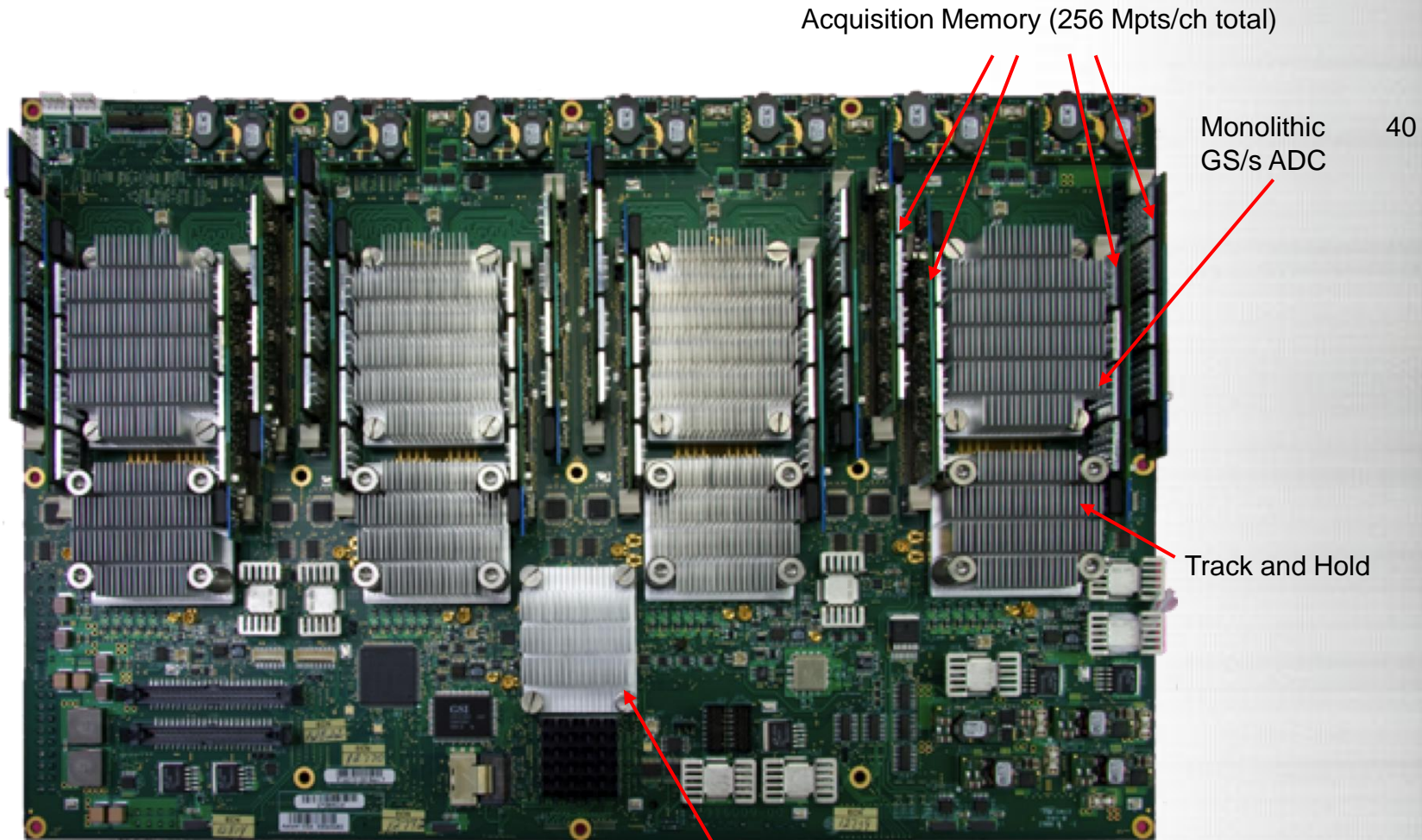
Real-Time Oscilloscopes

Front-end architecture



- High SR allows the acquired waveform samples to represent exactly the signal under test (Nyquist requirement)
- Track and Hold and A/D converters cover the entire bandwidth of the scope
- Trigger controls the storage of the digitized samples in memory

WaveMaster 8 Zi Acquisition Board



"Advanced Measurementsnot only Signal Integrity" - From DC to 30Ghz Real-Time Acquisition

Front End Architecture

Sample and Hold

- **“Sample-and-Hold” (MSH) front-end architecture to attain 16GHz Bandwidth**
 - Track and Hold circuits are placed immediately after high bandwidth front end amplifier.
 - ADCs can then digitized the tracked signals
 - MSH simplifies analog routing of high frequency signals
 - 15 GHz Edge Trigger chip
- **Monolithic 40GS/s ADC**

Digital Bandwidth Interleaving



(12) **United States Patent**
Pupalaikis et al.
 (10) **Patent No.:** US 7,058,548 B2
 (45) **Date of Patent:** Jun. 6, 2006

(54) **HIGH BANDWIDTH REAL-TIME OSCILLOSCOPE**

(75) Inventors: **Peter J. Pupalaikis**, Ramsey, NJ (US); **David C. Graef**, Campbell Hall, NY (US)

(73) Assignee: **LeCroy Corporation**, Chestnut Ridge, NY (US)

(*) Notice: Subject to any disclaimer, the term of this patent is extended or adjusted under 35 U.S.C. 154(b) by 0 days.

(21) Appl. No.: 10/693,188

(22) Filed: **Oct. 24, 2003**

(65) **Prior Publication Data**
 US 2004/0128076 A1 Jul. 1, 2004

Related U.S. Application Data

(60) Provisional application No. 60/420,937, filed on Oct. 24, 2002.

(51) **Int. Cl.**
G01R 23/00 (2006.01)

(52) **U.S. Cl.** **702/189; 702/75; 702/76; 341/155; 341/126**

(58) **Field of Classification Search** 702/189, 702/66, 67, 69-71, 73-76, 106, 112, 124, 702/126, 110, 190, 191, 195, 197, 198; 324/76.19, 324/76.22, 76.23, 76.24, 76.28, 76.29, 76.31, 324/76.38, 76.41-76.47; 327/91, 94, 100, 327/107, 129; 341/122, 123, 126, 155; 375/224, 375/225, 316; 708/300, 309, 311; 331/42, 331/43, 30-32, 64, 135; 382/260

See application file for complete search history.

(56) **References Cited**

U.S. PATENT DOCUMENTS

3,783,413 A	1/1974	Frommer et al.	
3,891,803 A	6/1975	Dagnet et al.	
3,903,484 A *	9/1975	Testani	331/135
4,316,282 A	2/1982	Macina	
5,659,546 A *	8/1997	Elder	370/343
5,668,836 A *	9/1997	Smith et al.	375/316
5,950,119 A *	9/1999	McGeehan et al.	455/302
6,240,150 B1	5/2001	Durveau et al.	
6,340,883 B1 *	1/2002	Nara et al.	324/76.78
2002/0150173 A1 *	10/2002	Buda	375/316
2004/0041599 A1 *	3/2004	Murphy	327/129

FOREIGN PATENT DOCUMENTS

EP	0 275 136	7/1988
EP	0 589 594	3/1994

OTHER PUBLICATIONS

Real-Time Spectrum Analysis Tools Aid Transition to Third-Generation Wireless Technology; Tektronix, Inc. 1999, pp. 1-6, no month.

(Continued)

Primary Examiner—Hal Wachsman
(74) Attorney, Agent, or Firm—Frommer Lawrence & Haug LLP; William S. Frommer

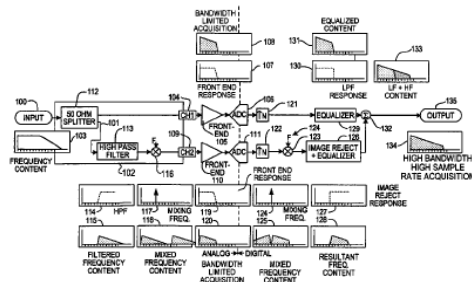
(57) **ABSTRACT**

A method and apparatus for digitizing a data signal. An input analog data signal, is received and split into a plurality of split signals. At least one of the split signals is mixed with a predetermined periodic function with a predetermined frequency. The split signals are then digitized and combined mathematically to form a single output data stream that is a substantially correct representation of the original input signal.

28 Claims, 23 Drawing Sheets

“DBI is a technology which combines the resources of multiple channels resulting in a channel with the sum of the individual source channels resources”.

- ✓ **Memory**
- ✓ **Sample Rate**
- ✓ **Bandwidth**

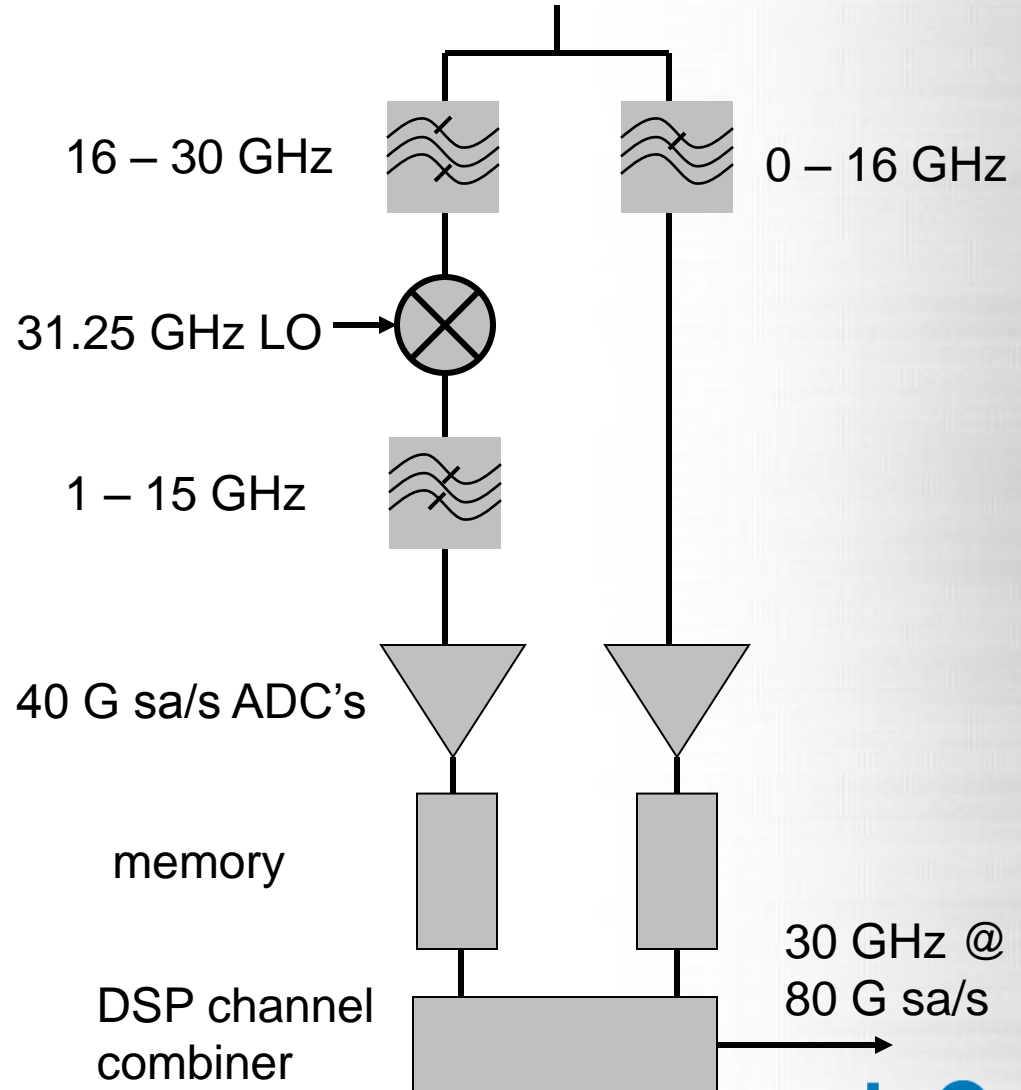


DBI Explained

Bandwidth Interleaving

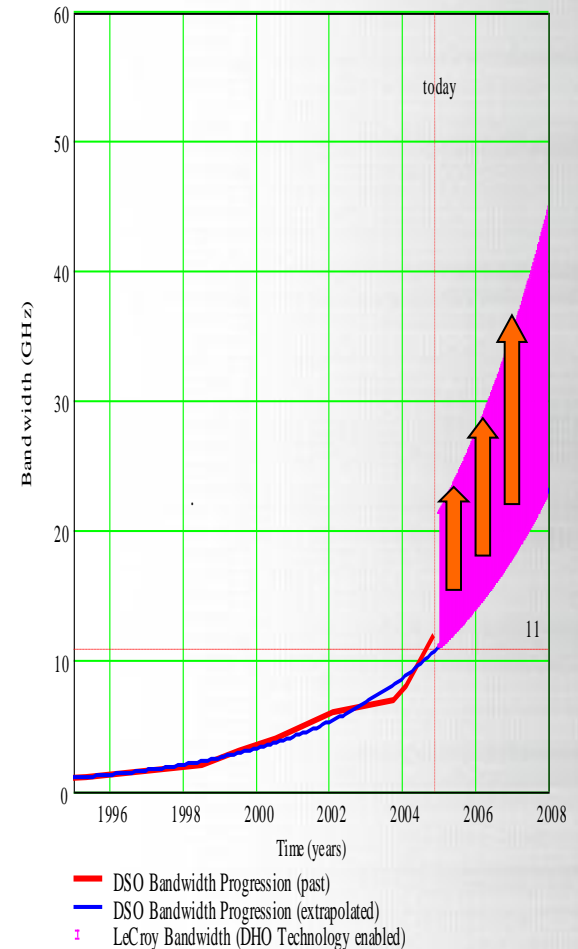
An innovative method to stay ahead of the BW curve

- **Separate signal into frequency bands**
- **Down-convert high band to low frequency**
- **Digitize down-converted and low-band simultaneously**
- **Use DSP to compensate delay, phase and amplitude and combine bands**



DBI – A Whole New Ballgame

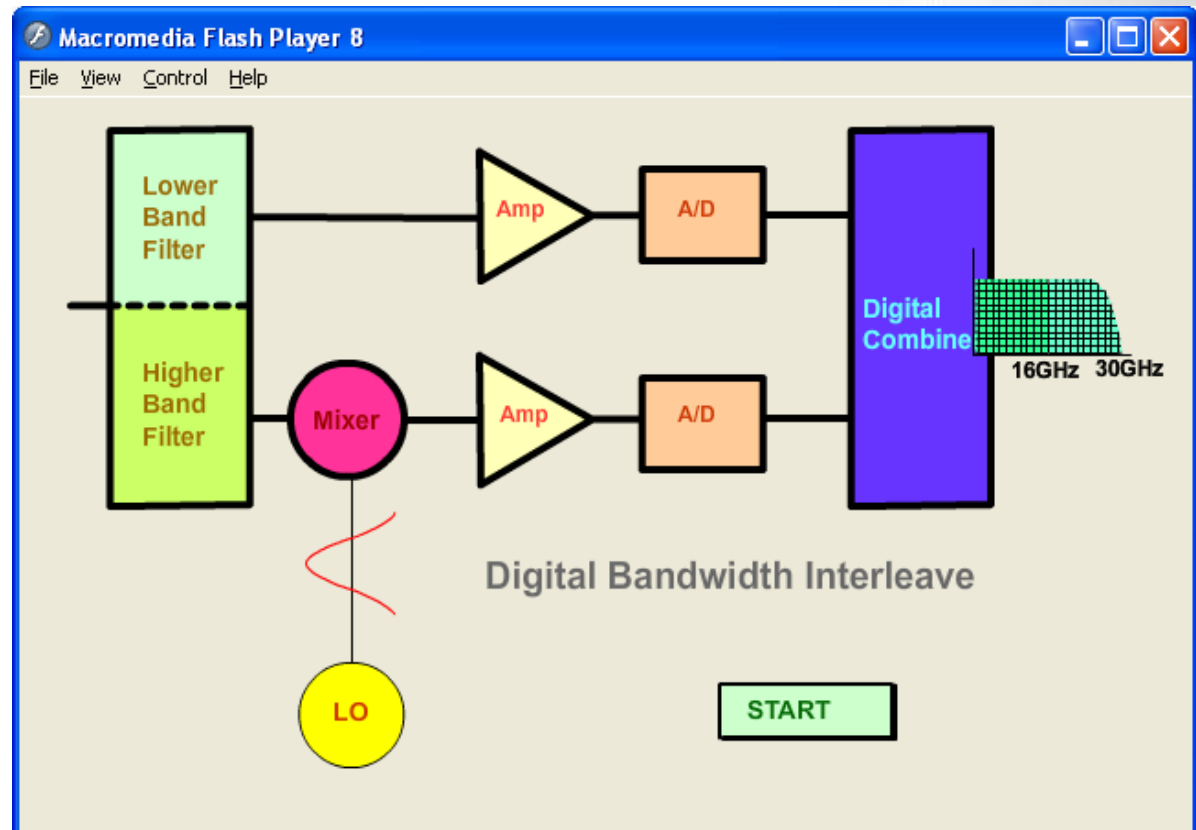
- ✓ **Moore's Law:**
 - ✓ *Transistors density continues to double every 18 months; the speed of circuit is doubling every three years.*
- ✓ **Test Equipment manufacturers use the same technology employed by state of the art devices under test**
- ✓ **DBI is a new way to extend the bandwidth and the sample rate of a silicon platform**
- ✓ **i.e. a step function to the Moore's Law**



Bandwidth Interleaving - Animation



DBI Animation -
Click Here



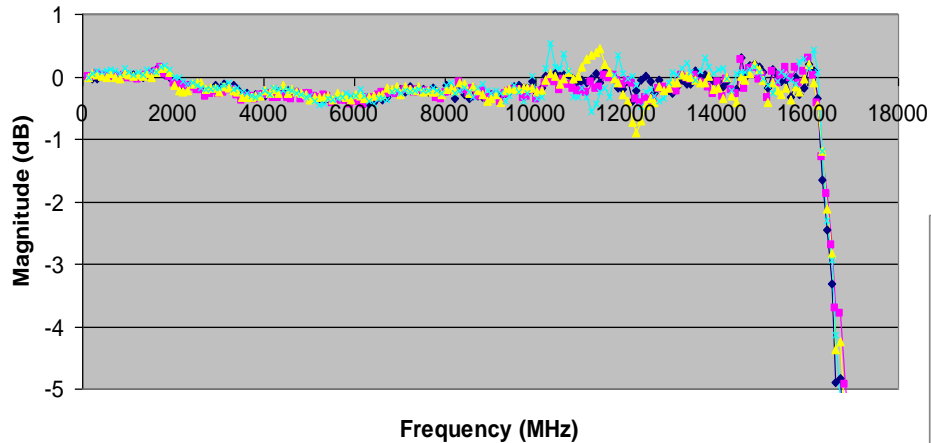
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Frequency Response Optimization

LeCroy Real-Time 7 Zi and 8 Zi platforms

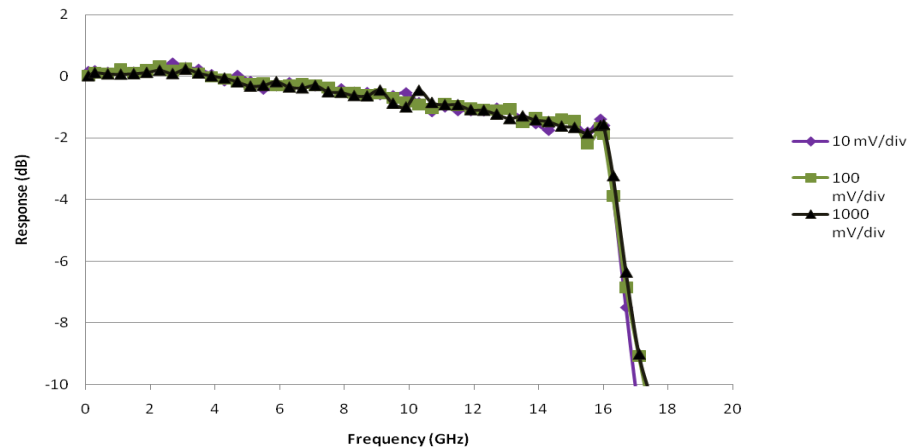
LeCroy WM 816 Zi 16GHz Mode, Flatness Optimization



Flatness Optimization Mode



WaveMaster 816Zi Bandwidth Response
16 GHz, Ch1, Pulse/Eye Optimization Mode



Pulse/Eye Optimization Mode



Frequency Response (i.e. Magnitude Response roll-off and Phase Response) influences rise time and the overshoot of the step response in time domain.

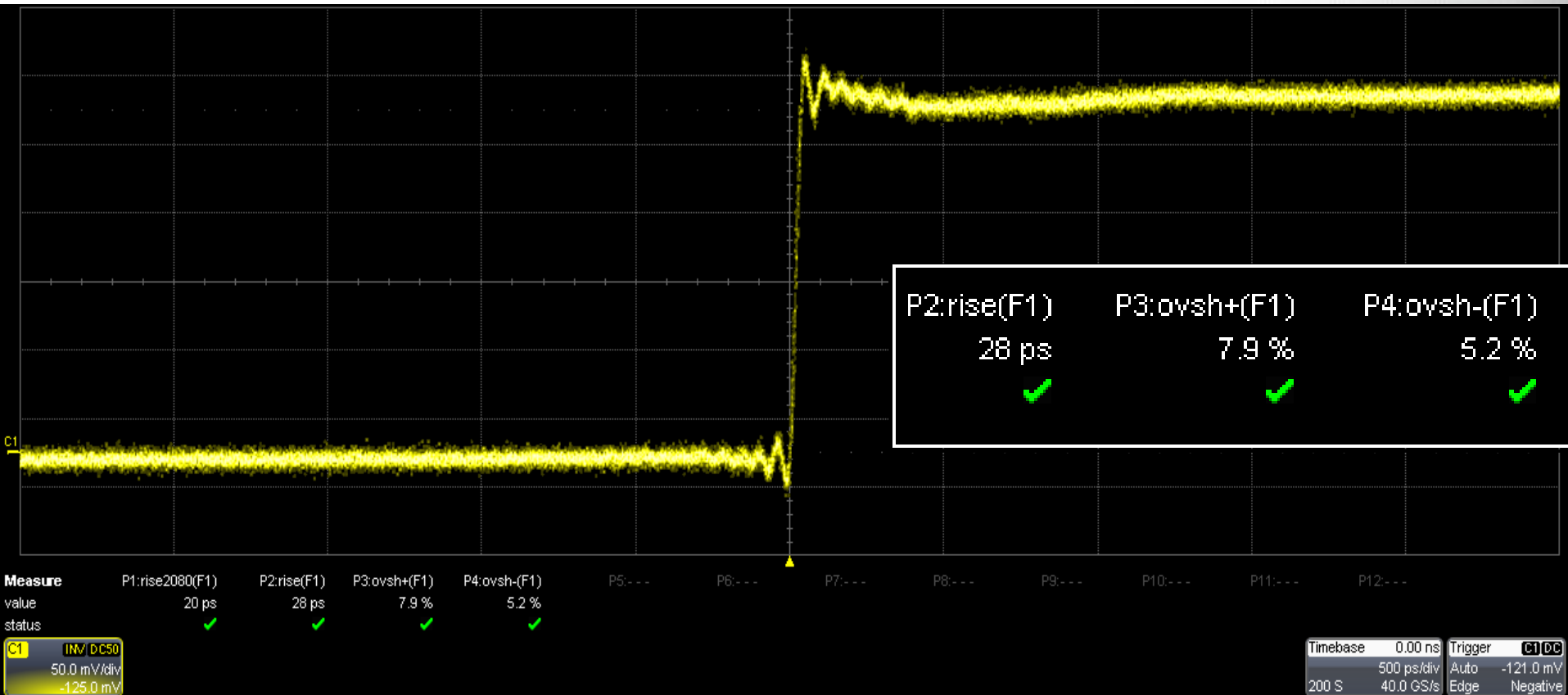
A “brick wall” rolloff produces higher overshoot and shorter rise time.

A “slow” rolloff produces less overshoot but longer rise time.

User can optimize frequency response to his application

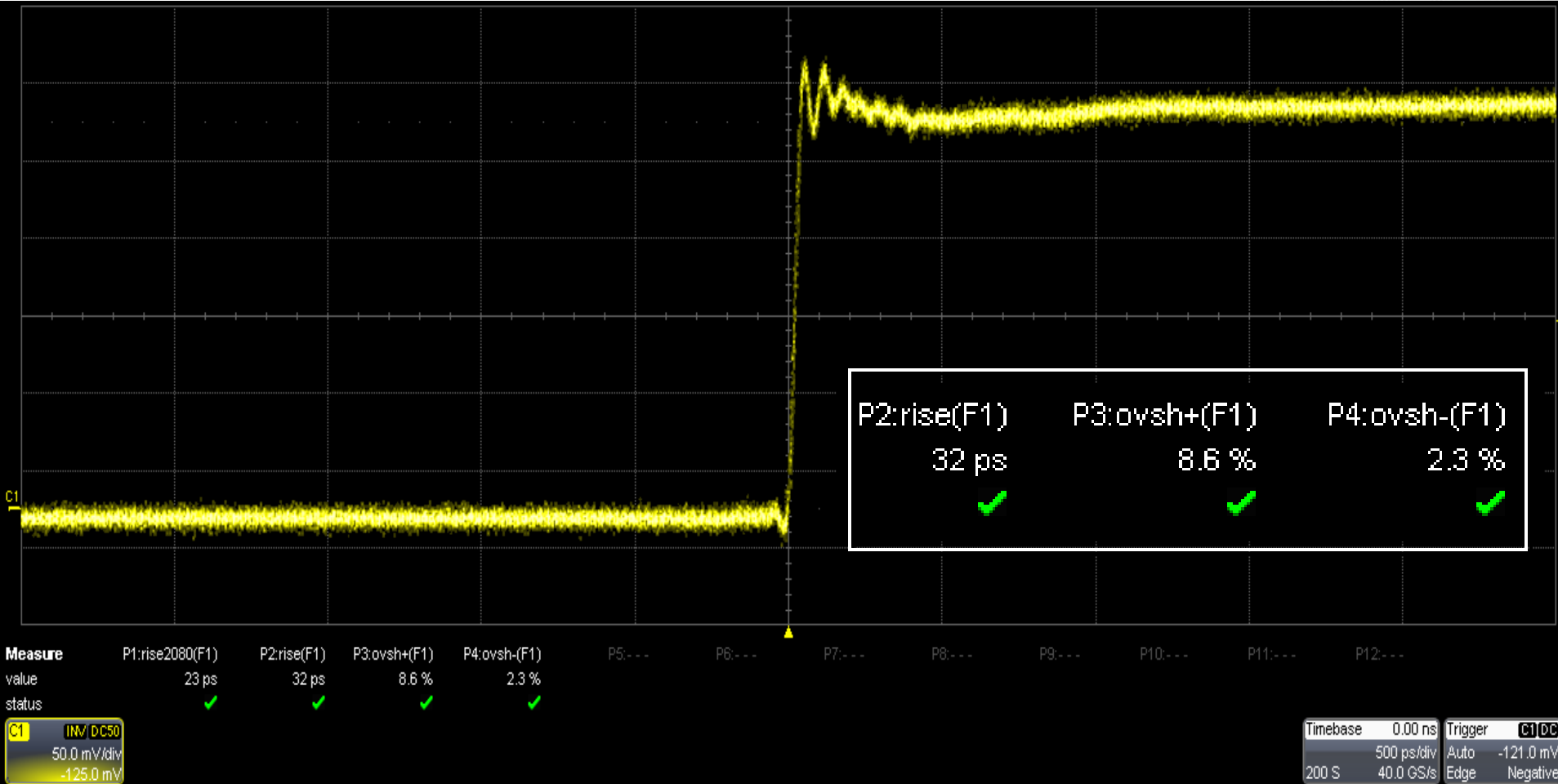
Optimization Mode	Magnitude Response	Phase Response "Group Delay"	Benefits	Tradeoff	Typical Application
Pulse Response	4 th order Bessel-Thompson	Minimum Phase	Closely mimics an analog scope.	Slower edge speed	General purpose
Eye Diagram	4 th order Bessel-Thompson	Linear Phase	Linear Phase Response reduces overshoot and improves edge speed.	Increased preshoot since preshoot and overshoot are equalized (Doesn't matter for serial data)	Serial Data Testing
Flatness	Brickwall	Linear Phase	Maximum flatness in the pass band.	Increased Overshoot (Doesn't matter for sine waves or modulation)	Spectral Analysis

“Flatness Mode” Optimization

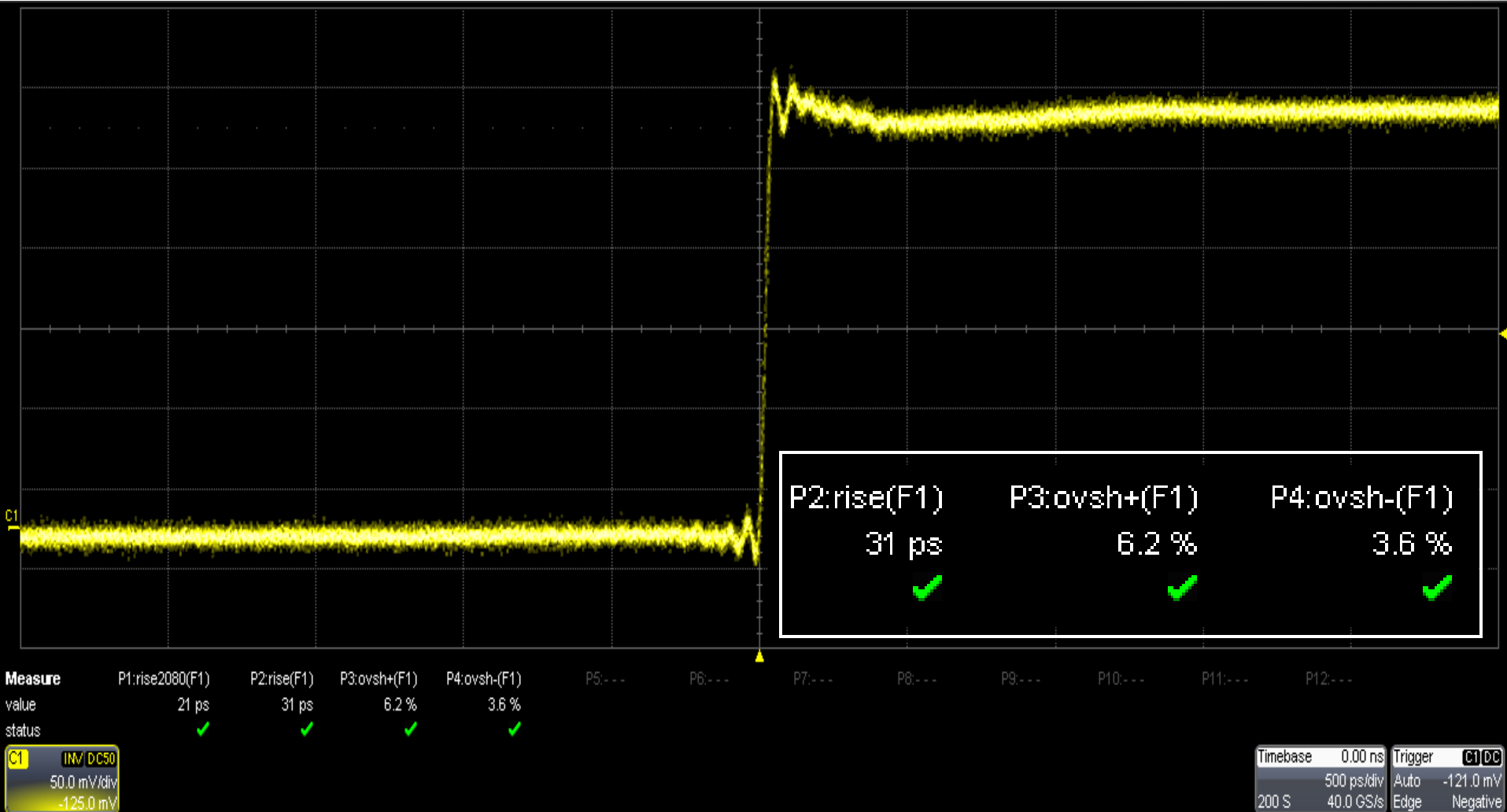


$$\text{Scope Only Risetime} = \sqrt{\text{Measured Risetime}^2 - \text{Input Risetime}^2}$$

“Pulse Response Mode” Optimization



“Eye Diagram Mode” Optimization



Frequency Response Optimization

Example : Real-Time 16GHz Oscilloscope

	PULSE RESPONSE	EYE DIAGRAM	FLATNESS
Measured Risetime (10%-90%)	29ps	27ps	25ps
Preshoot	2%	6%	8%
Overshoot	13%	6%	8%
Magnitude Response Bandwidth Shape	Bessel	Bessel	Flat
Phase Response	Minimum	Linear	Linear

PulseResponse

EyeDiagram

Flatness

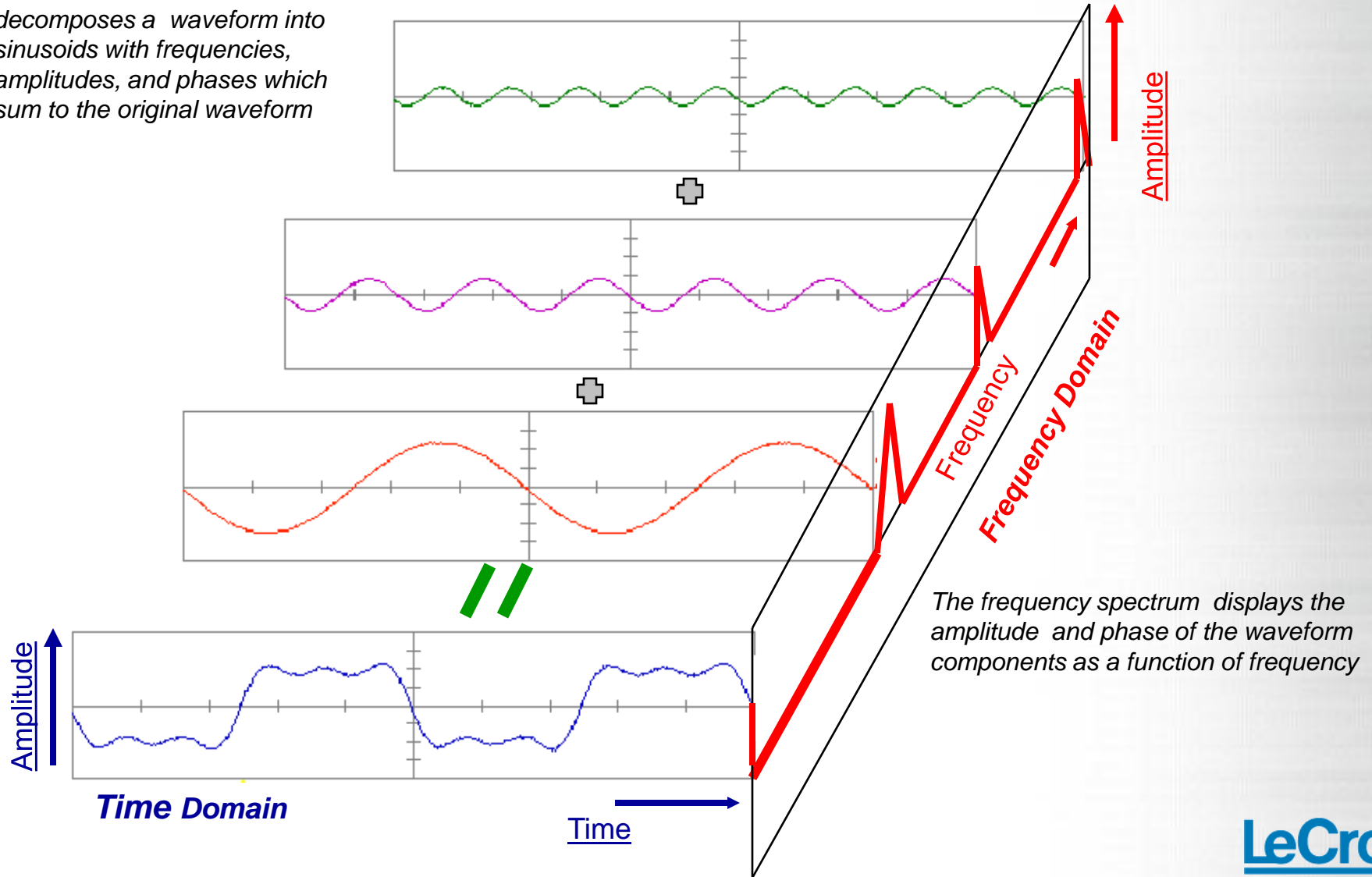
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Time and Frequency Domain

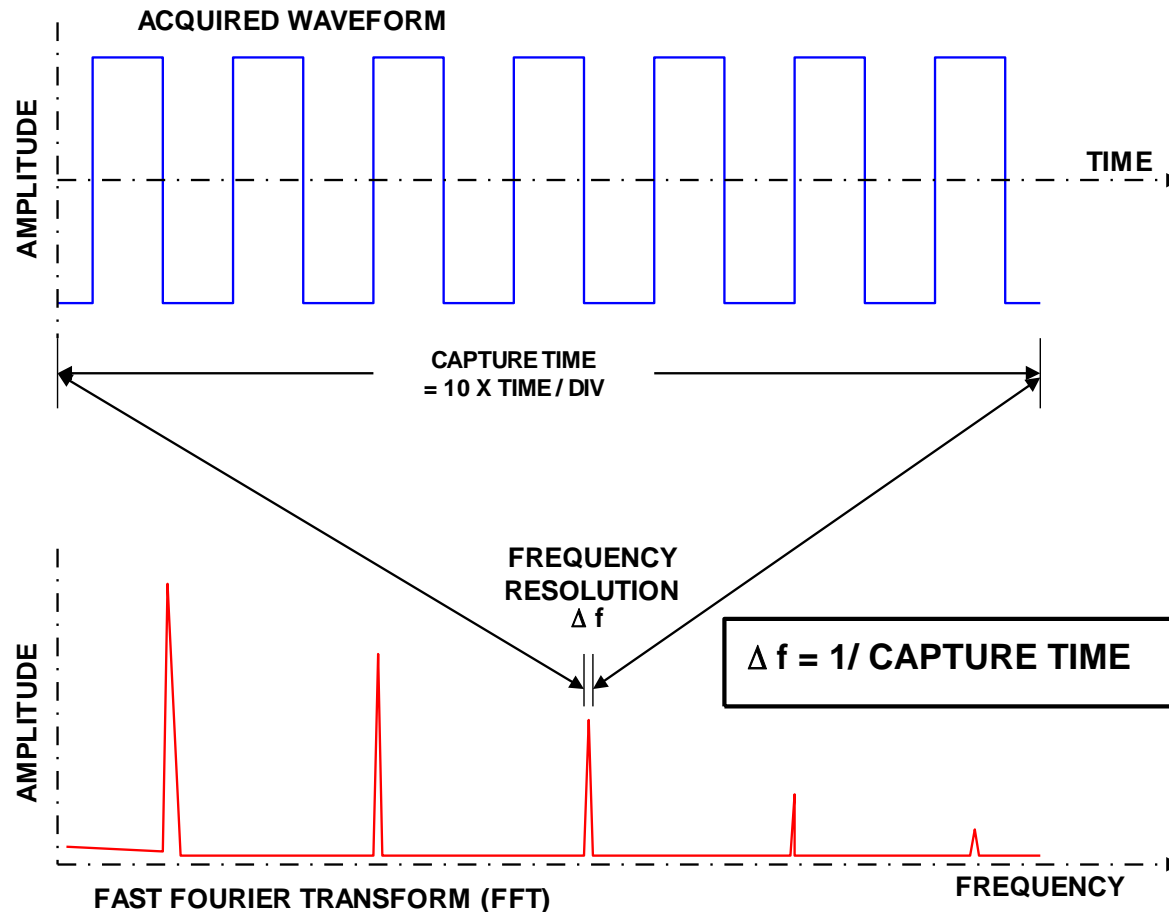
FFT Basics

The Fourier transform decomposes a waveform into sinusoids with frequencies, amplitudes, and phases which sum to the original waveform



Frequency Resolution, Δf

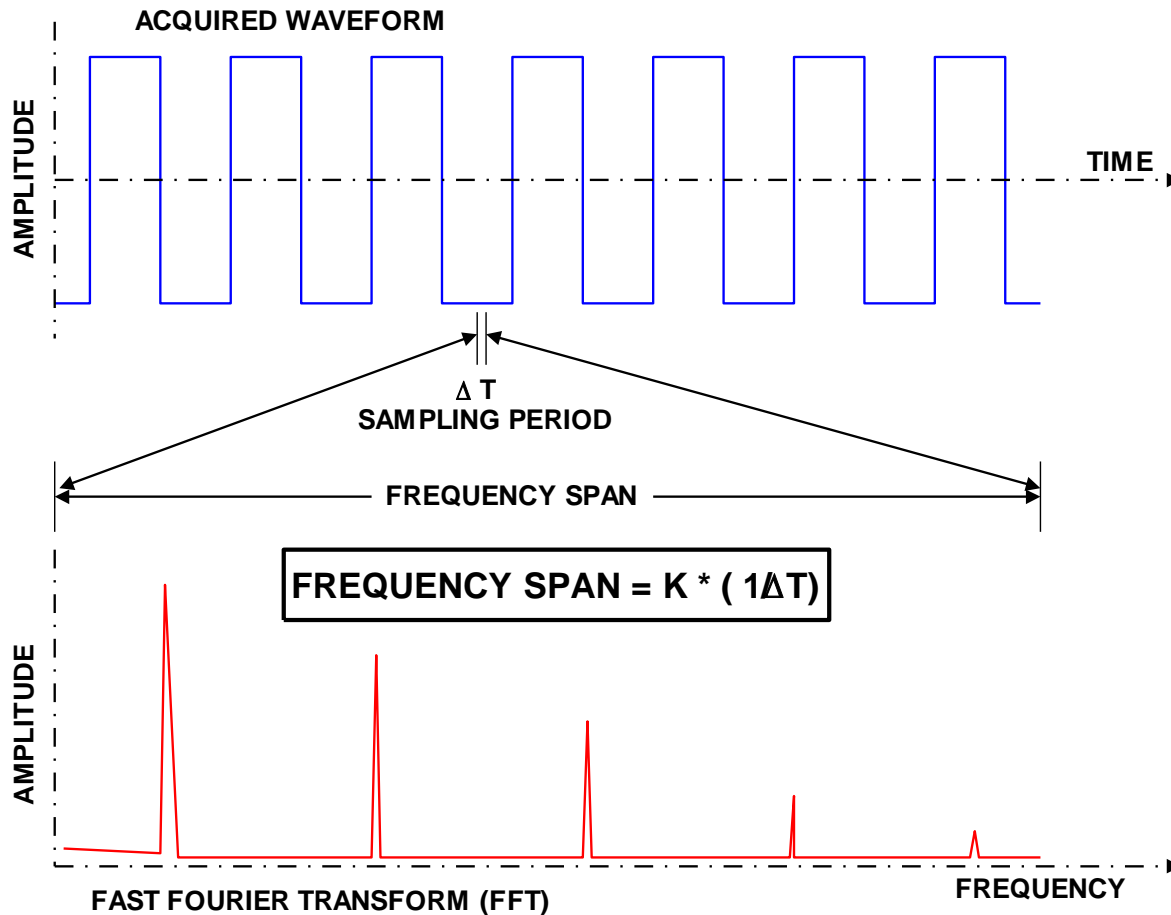
FFT Basics



- The frequency spectrum of an FFT is quantized with a step size, Δf , defined by the resolution bandwidth
- The resolution bandwidth of the FFT is the reciprocal of the capture time
- The greater the required frequency resolution (smaller Δf) the longer the record length must be
- A resolution bandwidth of 1Hz requires a 1 second record length

Frequency Span Range

FFT Basics

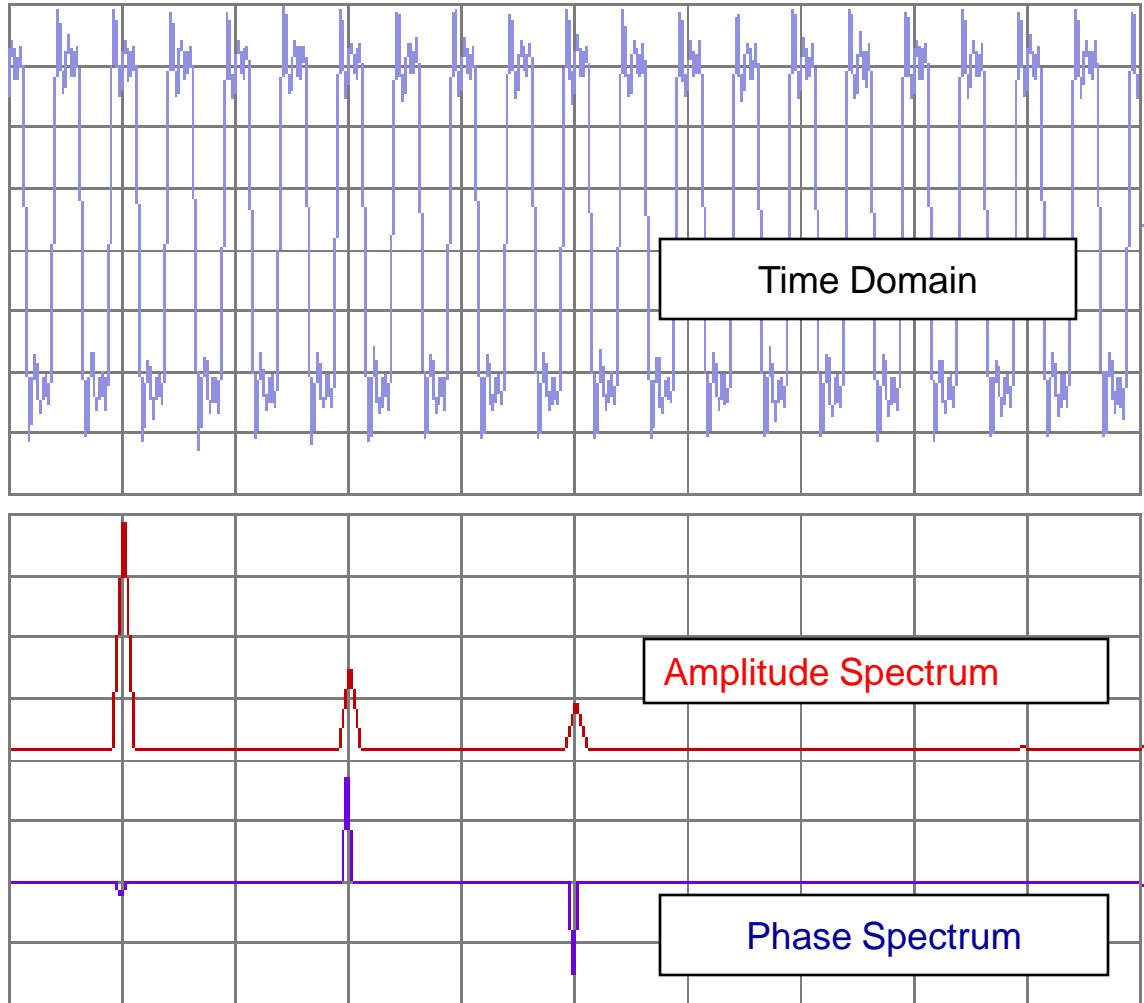


- The frequency span, or Nyquist frequency, of the FFT is proportional to 1/2 the effective sampling rate
- The effective sampling rate is the acquisition sampling rate
- The Sparsing math function can be applied to reduce the FFT span
- The effective sampling rate with sparsing is the acquisition rate divided by the sparsing factor

The Frequency Span or Nyquist frequency is proportional to the effective acquisition sampling rate

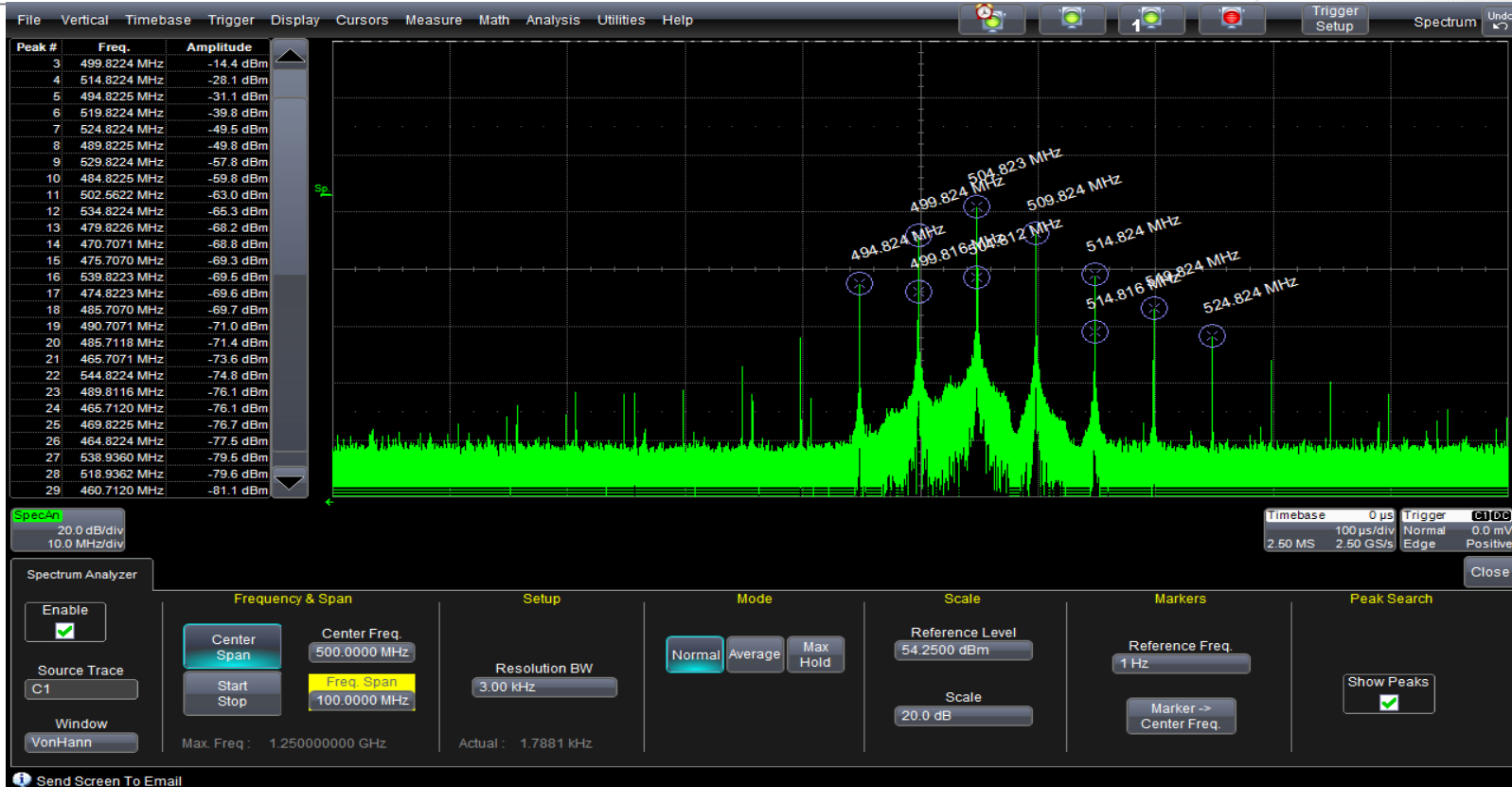
Magnitude And Phase Spectra

FFT Basics



- The frequency domain output of the Fourier transform is complex
- A complete description must include: Magnitude and Phase OR Real and Imaginary components
- The magnitude spectrum can be displayed in linear units or units of power
- Vertical scale formats can be linear or logarithmic

SPECTRUM Analyzer Option Simplifies Frequency Analysis, Adds Capability

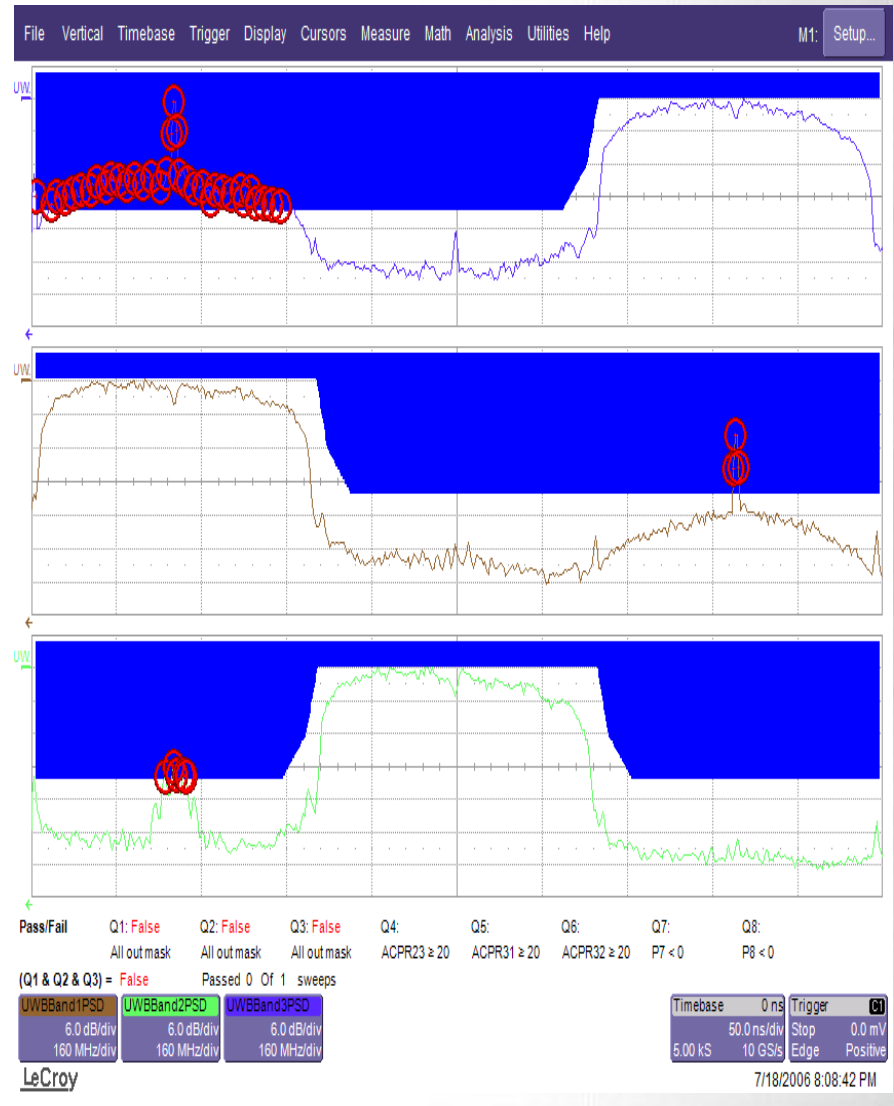
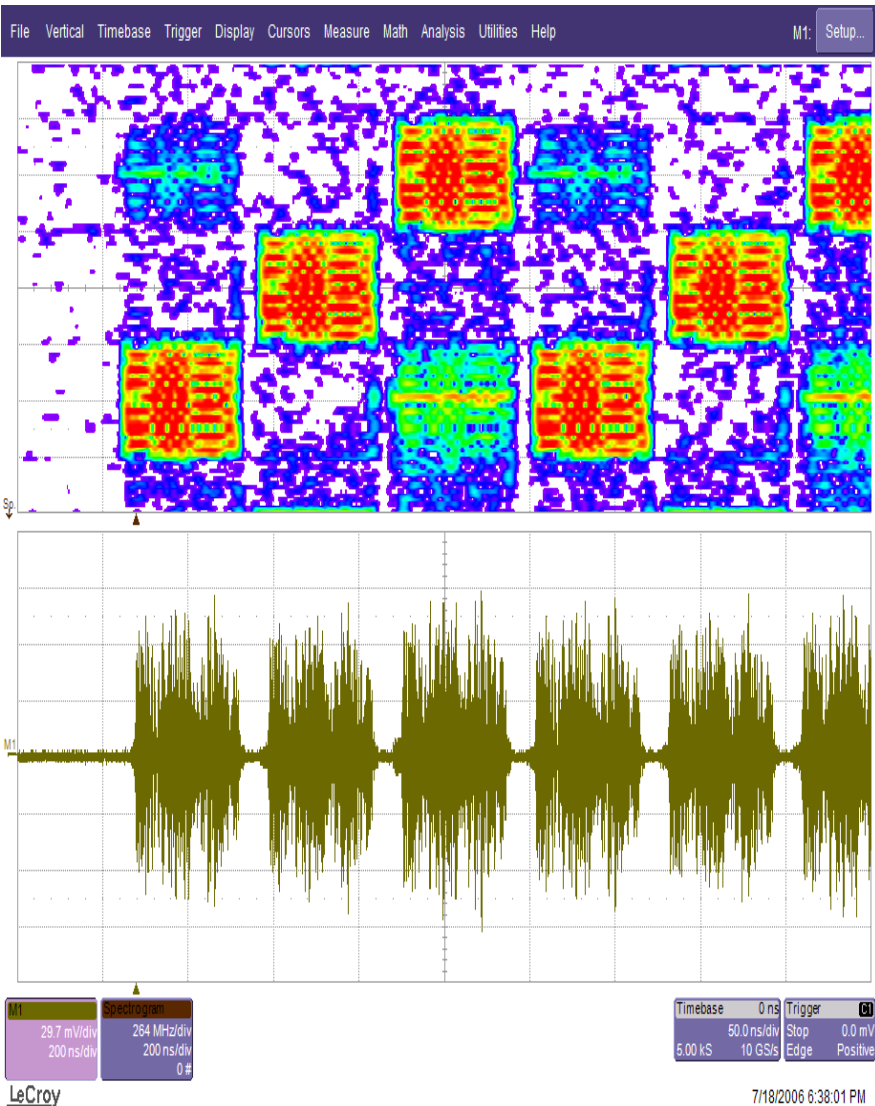


Common User Interface as RF Spectrum Analyzer :

- Center Frequency, Span, Resolution BW
- Averaging and Max Hold, Ref level
- Find Peaks capability

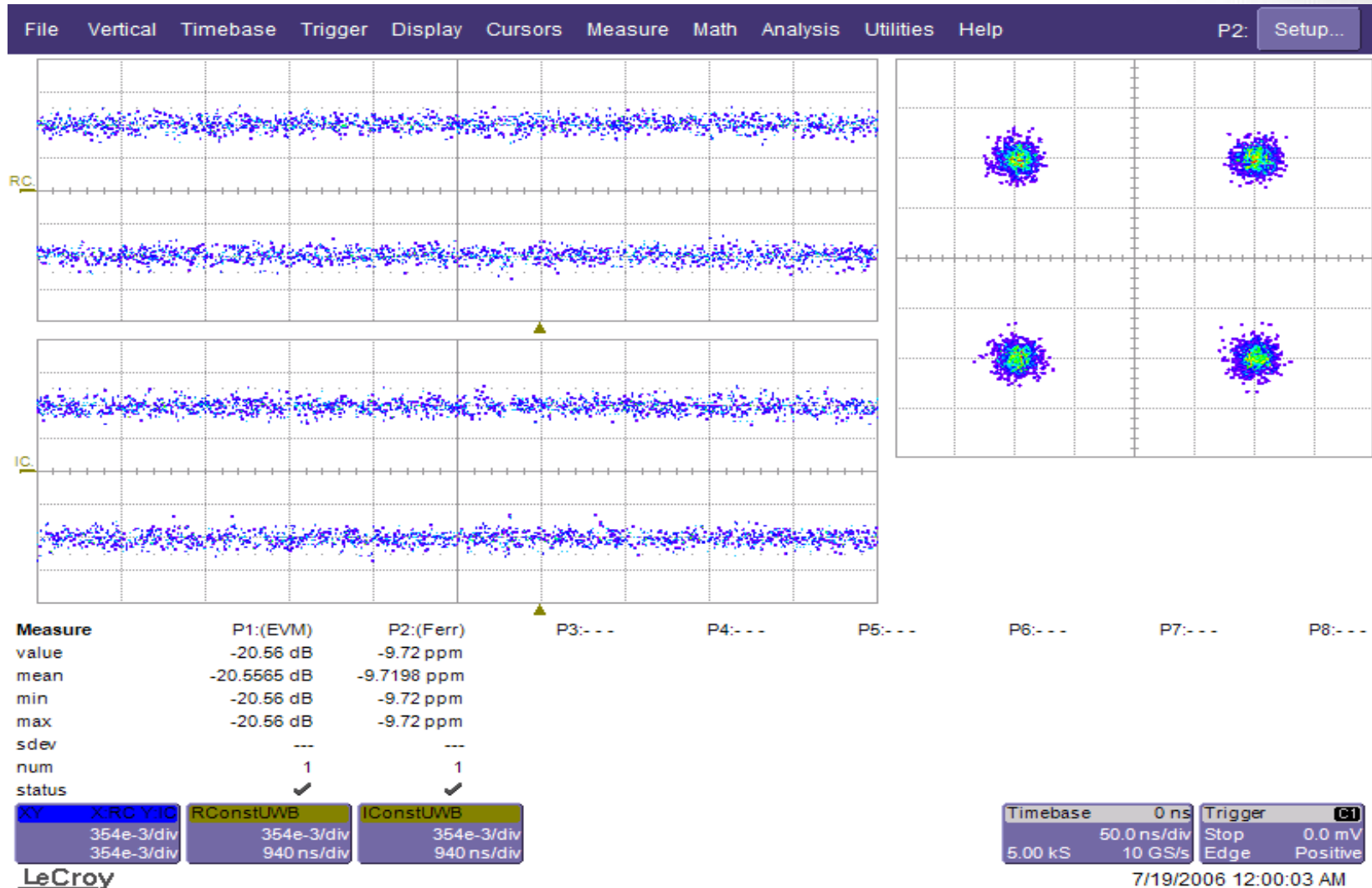
Spectral display with mask test

Example of UWB with TFC 1 packet



EVM / Constellation display

Example of UWB with TFC 1 packet



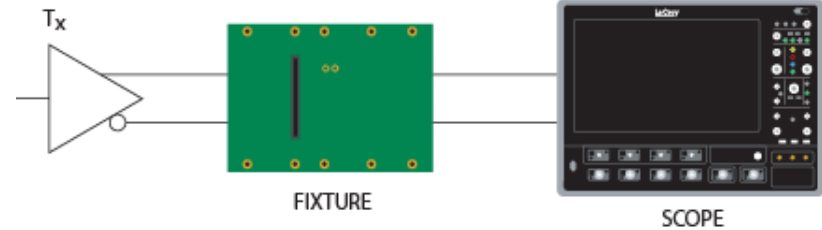
LeCroy

Agenda

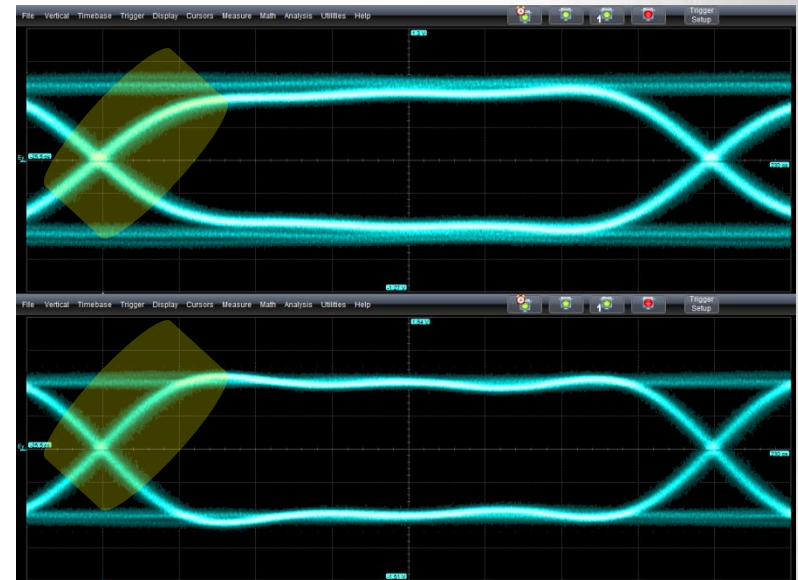
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Cable/Fixture/Serial Data Channel De-embedding

- ✓ *De-embedding gives the user the ability to view their waveforms as if the cable/fixture/serial data channel was not present.*
- ✓ *Cable/fixture/serial data channel effect is defined by user-supplied S-parameter file.*
- ✓ *Should be always done when frequency content is increasing into the microwave range*

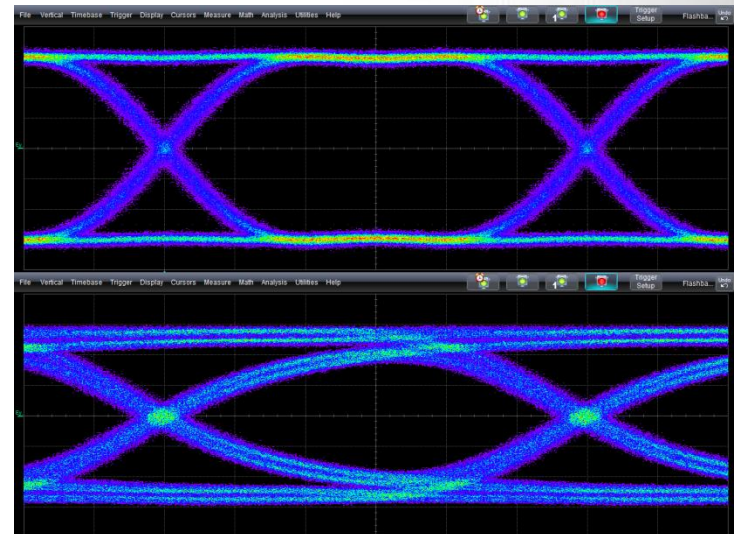
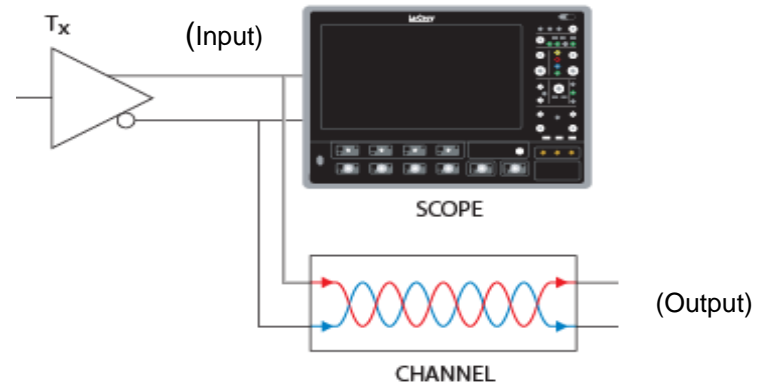


Cables and fixture de-embedding effects on a PCI's Gen 2 signal



Serial Data Channel Response Emulation

- ✓ *Channel emulation gives the user the ability to view their waveform as if it were passing through a serial data channel that isn't present.*
- ✓ *The channel effect is defined by a user-supplied s-parameter file.*
- ✓ *Useful to model worst-case scenarios and understand how design margins are maintained in those situations*



Sampling and Real-Time good companion for Signal Integrity Measurements

